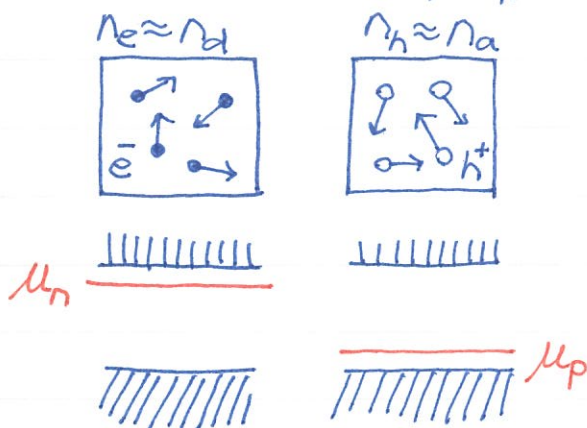


## HH0042 p-n Junction

When a semiconductor is NOT uniformly doped, interesting transport properties show up and are of crucial importance for device applications. Here we just demonstrate the simple p-n junction by putting p-type and n-type semiconductors together with sharp interface.

① **Qualitative understanding:** Consider two pieces of extrinsic semiconductors: p-type and n-type. As discussed before, the chemical potentials  $\mu_n, \mu_p$  may not be the same. For simplicity, we assume complete ionization so that



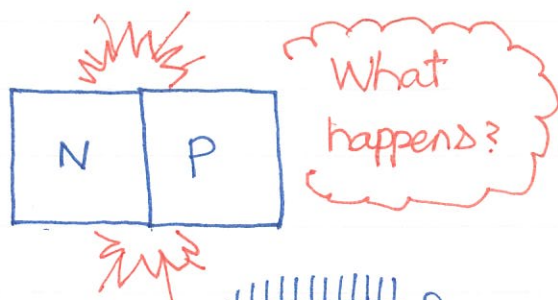
chemical potentials  $\mu_n, \mu_p$  may not be the same. For simplicity, we assume complete ionization so that

$$n_e \approx n_d^+ \approx n_d \text{ (n-type)}$$

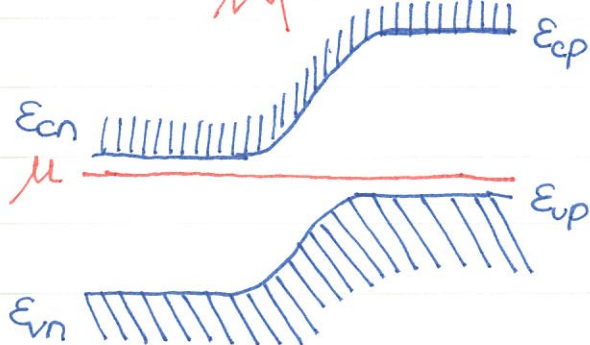
electrons are majority.

$$n_h \approx n_a^- \approx n_a \text{ (p-type)}$$

holes are majority.



After carrier migration, there is just ONE chemical potential. Thus, one can guess the bands are shifted...



$$E_c(x) = E_c(-\infty) - e\phi(x)$$

with  $\phi(-\infty) = 0$  and  $E_c(-\infty) = E_{cn}$

From the figure on the left, it's clear that  $E_c(\infty) = E_{cp}$

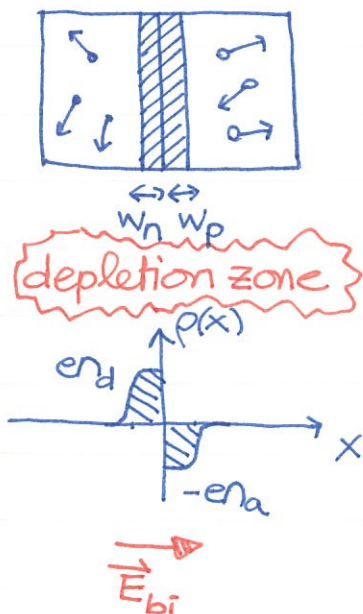
For the valence band edge,  $E_v(x) = E_v(-\infty) - e\phi(x)$  where  $E_v(-\infty) = E_{vn}$

Similarly, it's easy to see that  $E_v(\infty) = E_{vp}$ . The band gap is assumed to be robust  $E_g \equiv E_c(x) - E_v(x)$  and a built-in voltage

$$eV_{bi} = E_{cp} - E_{cn} = E_{vp} - E_{vn}$$

is present due to charge migration to

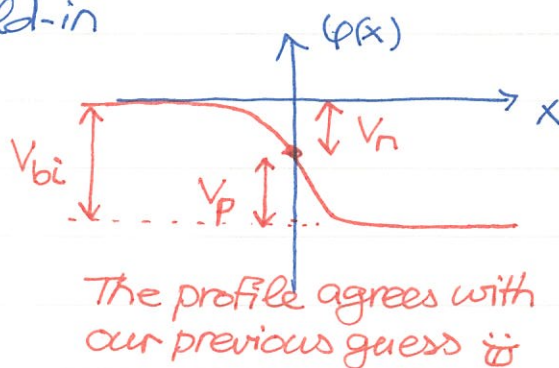
Simple picture



Suppose the interface is at the origin  $x=0$ . The holes in the p-type migrate over the interface and annihilate with electrons. Similarly, the electrons in the n-type migrate and annihilate with holes on the other side.  $\rightarrow$  formation of depletion zone! Inside the depletion zone, there is almost no mobile electrons and holes

$$\rho(x) \approx \begin{cases} e n_d & \text{on the n-type side} \\ -e n_a & \text{on the p-type side} \end{cases}$$

The charges give rise to a built-in electric field  $\vec{E}_{bi}$ , pointing from n-type to p-type and stopping further charge migration from both sides. The built-in electric field generates a potential profile  $\phi(x)$  as shown on the right.  $\phi(-\infty) \equiv 0$  as the zero for the electrostatic potential.



① Finding build-in voltage: We assume the extrinsic S-C. is in the regime satisfying (1)  $n_i \ll n_d, n_a \ll n_c, n_v$  ← "classical"  
 (2)  $n_e \approx n_d, n_h \approx n_a$  ← fully ionized!

So, starting from the n-type side,  $n_e = n_c e^{-(E_{cn} - \mu)/\tau} \rightarrow E_{cn} = \mu + \tau \log\left(\frac{n_c}{n_e}\right)$

Making use of  $n_e \approx n_d \rightarrow E_{cn} = \mu + \tau \log\left(\frac{n_c}{n_d}\right)$

Now turn to the p-type side.

$n_h = n_v e^{-(\mu - E_{vp})/\tau} \rightarrow E_{vp} = \mu - \tau \log\left(\frac{n_v}{n_h}\right) = \mu - \tau \log\left(\frac{n_v}{n_a}\right)$

Thus,  $E_{cn} - E_{vp} = \tau \log\left(\frac{n_c n_v}{n_d n_a}\right)$  we are almost there ☺

From the band-bending diagram,  $E_{vp} = E_{vn} + eV_{bi}$

$$E_{cn} - E_{vp} = E_{cn} - E_{vn} - eV_{bi} = \underbrace{E_g}_{\text{band gap}} - eV_{bi}, \quad E_g \text{ is the band gap.}$$

Finally, it is straightforward to solve for the built-in voltage.

$$eV_{bi} = E_g - \tau \log\left(\frac{n_c n_v}{n_d n_a}\right) < E_g$$

Note that  $eV_{bi}$  is of the same order of  $E_g$ , but slightly smaller.

① The tougher part - finding  $\psi(x)$ : To find the profile of the electrostatic potential, one needs to solve the Poisson eq. self-consistently

$$\frac{d^2\psi}{dx^2} = -\frac{1}{\epsilon} \rho$$

$\rho = \rho[\psi]$  depends on  $\psi(x)$  ..... tough  $\frac{d^2}{dx^2}$ !!!

Again, one needs to carry out the calculations on two sides. On the n-type side,  $\rho(x) = e[n_d - n_e(x)]$ . w/  $n_e(x) = n_c e^{-(E_c(x) - \mu)/\tau}$

$$\rightarrow n_e(x) = \underbrace{n_c e^{-(E_{cn} - \mu)/\tau}}_{= n_e(-\infty) \approx n_d} e^{e\psi(x)/\tau} = \underbrace{n_d e^{e\psi(x)/\tau}}_{\text{depending on } \psi(x)!}$$

Now we can write down the Poisson eq. on the n-type side,

$$\frac{d^2\psi}{dx^2} = -\frac{en_d}{\epsilon} [1 - e^{e\psi(x)/\tau}] < 0$$

Must be solved self consistently!

On the p-type side,  $n_h(x) = n_v e^{-[\mu - E_v(x)]/\tau}$  w/  $E_v(x) = E_v(-\infty) - e\psi(x)$

Because  $E_v(-\infty) = E_{vn} = E_{vp} - eV_{bi}$ , the hole concentration becomes

$$n_h(x) = \underbrace{n_v e^{-(\mu - E_{vp})/\tau}}_{\dots\dots\dots} e^{-e[\psi(x) + V_{bi}]} \approx n_a e^{-e[\psi(x) + V_{bi}]}$$

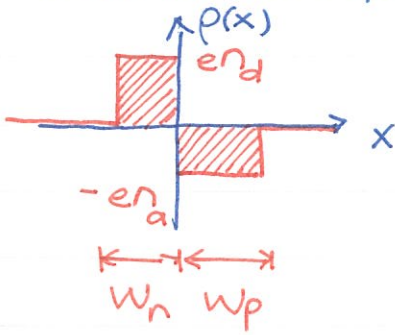
Following similar steps, the Poisson eq. now takes the form,

$$\frac{d^2\psi}{dx^2} = \frac{en_a}{\epsilon} [1 - e^{-e[\psi(x) + V_{bi}]}] > 0$$

$\frac{d^2\psi}{dx^2}$  has opposite signs

Without computers, it is very messy to solve the above equations self consistently. Thus, we are going to simplify them ....  $\infty$

Assume the depletion zone has the step-like charge distribution.



The approximation actually makes sense. On the n-type side,  $\rho(x) = en_d [1 - e^{e\phi(x)/kT}]$

$$\rightarrow \rho(x) \cong \begin{cases} en_d, & -w_n < x < 0 \\ 0, & x < -w_n \end{cases}$$

B.C.:  $\phi(-\infty) = 0$

The Poisson equation in the regime  $-w_n < x < 0$  is

$$\frac{d^2\phi_2}{dx^2} = -\frac{en_d}{\epsilon}$$

SO SIMPLE !!

$$\phi_2(x) = \begin{cases} -\frac{1}{2} \frac{en_d}{\epsilon} (x+w_n)^2 & -w_n < x < 0 \\ 0 & x < -w_n \end{cases}$$

On the p-type side,  $\rho(x) = -en_a [1 - e^{-e\phi(x) - eV_{bi}}]$

$$\rightarrow \rho(x) \cong \begin{cases} -en_a, & 0 < x < w_p \\ 0, & x > w_p \end{cases}$$

The Poisson equation also takes a very simple form

in the depletion zone  $0 < x < w_p$ :

$$\frac{d^2\phi_3}{dx^2} = \frac{en_a}{\epsilon}$$

SIMPLE !!

$$\phi_3(x) = \begin{cases} \frac{1}{2} \frac{en_a}{\epsilon} (x-w_p)^2 - V_{bi}, & 0 < x < w_p \\ -V_{bi}, & x > w_p \end{cases}$$

B.C.:  $\phi(\infty) = -V_{bi}$

But, we are NOT done yet.....

The depletion width  $w_n, w_p$  can be solved by matching B.C. at the interface  $x=0$ .

$$(1) \phi_2(0) = \phi_3(0) \quad -\frac{1}{2} \frac{en_d}{\epsilon} w_n^2 = \frac{1}{2} \frac{en_a}{\epsilon} w_p^2 - V_{bi}$$

$$\rightarrow \frac{1}{2} \frac{en_d}{\epsilon} w_n^2 + \frac{1}{2} \frac{en_a}{\epsilon} w_p^2 = V_{bi} \quad \leftarrow 1^{st} \text{ equation for } (w_n, w_p).$$

(2)  $\frac{d\phi_2}{dx}(0) = \frac{d\phi_3}{dx}(0)$  because there is no surface charge. So the electric field at the origin is well defined.

$$E(0) = -\frac{d\phi}{dx}(0) = \frac{en_d}{\epsilon} w_n = \frac{en_a}{\epsilon} w_p \quad \leftarrow 2^{nd} \text{ equation for } (w_n, w_p).$$

To compare with the results in Kittel, we can solve  $E(0)$  first.

$$\frac{1}{2} \frac{\epsilon}{\epsilon n_d} \left( \frac{\epsilon n_d}{\epsilon} W_n \right)^2 + \frac{1}{2} \frac{\epsilon}{\epsilon n_a} \left( \frac{\epsilon n_a}{\epsilon} W_p \right)^2 = V_{bi} \quad \text{--- } E(0)$$

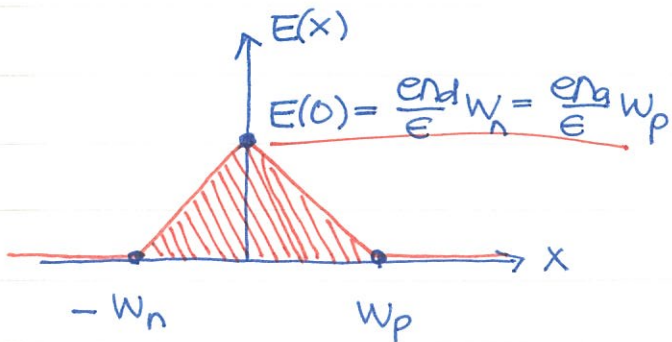
One can find the electric field at the interface  $x=0$ :

$$\frac{\epsilon}{2e} \left( \frac{1}{n_d} + \frac{1}{n_a} \right) E(0)^2 = V_{bi} \quad \rightarrow \quad E(0) = \sqrt{\frac{2e}{\epsilon} \frac{n_a n_d}{n_a + n_d} V_{bi}}$$

The above result is basically the same as in Kittel's textbook. Now you know the physical picture behind the approximation made in the textbook. Finding the depletion length  $W_n, W_p$  is now quite trivial.

$$W_n = \frac{\epsilon}{\epsilon n_d} E(0) \quad , \quad W_p = \frac{\epsilon}{\epsilon n_a} E(0)$$

I shall not bore you to write down the expressions explicitly. It's helpful to plot  $E(x)$  within this approximation,



The area under the  $E(x)$  curve is the built-in voltage

$$\frac{1}{2} \frac{\epsilon n_d}{\epsilon} W_n^2 + \frac{1}{2} \frac{\epsilon n_a}{\epsilon} W_p^2 = V_{bi}$$

The linear dependence of  $E(x)$  explains why the depletion length  $W_n, W_p \propto \sqrt{V_{bi}}$  as derived previously.



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