

# CHAPTER 2



## CIRCUIT ELEMENTS AND KIRCHHOFF'S LAWS

# CONTENTS

2.1 Active Components

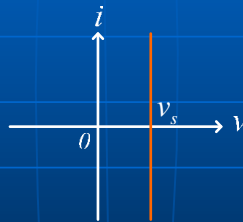
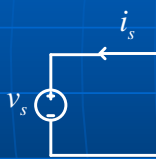
2.2 Passive Components

2.3 Node , Branch and Loop

2.4 Kirchhoff's Laws

## 2.1 Active Components

- Ideal independent voltage source
- Symbol and model



$$i_s = \infty$$

$$i_s = 100 \text{ A}$$

$$i_s = 0 \text{ A}$$

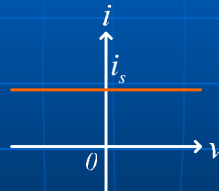
$$i_s = -50 \text{ A}$$

$$i_s = -\infty$$

- It maintains a prescribed voltage independent of the current through this source.

## 2.1 Active Components

- Ideal independent current source
- Symbol and model



$$v_s = \infty$$

$$v_s = 100 \text{ V}$$

$$v_s = 0 \text{ V}$$

$$v_s = -50 \text{ V}$$

$$v_s = -\infty$$

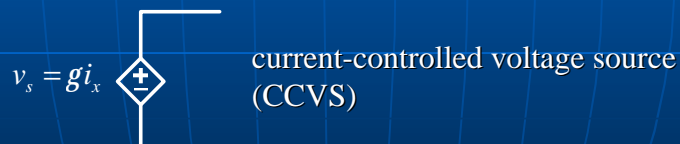
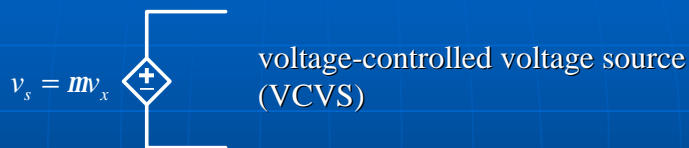
- It maintains a prescribed current independent of the voltage across this source.

## 2.1 Active Components

- n Ideal dependent voltage and current sources are convenient for modeling transistors
- n An ideal dependent (or controlled) voltage source is an active element whose source quantity is controlled by other voltage or current.

## 2.1 Active Components

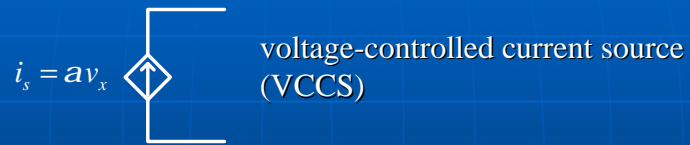
- n Symbol and model of dependent voltage source



$v_x$  and  $i_x$  are controlling parameter.

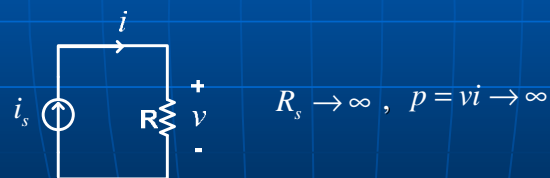
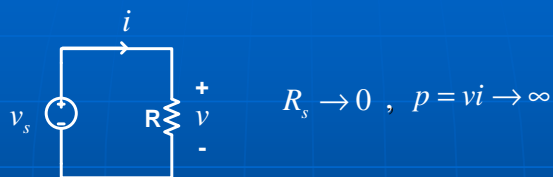
## 2.1 Active Components

n Symbol and model of dependent current source



$v_x$  and  $i_x$  are controlling parameter.

## 2.1 Active Components

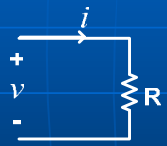


Ideal sources do not exist in real world.

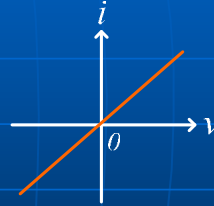
## 2.2 Passive Components

### (a) Resistance

n Symbol and model



$$v = Ri$$
$$i = Gv$$

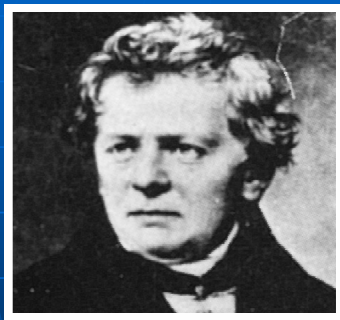


R: resistance , in ohms ( $\Omega$ )

G: conductance , in siemens (S)

Ohm's Law

## 2.2 Passive Components



Georg Simon Ohm  
1787-1854  
Germany

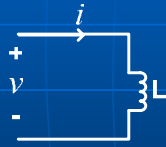


???  
19XX-20XX  
R.O.C. Taiwan

## 2.2 Passive Components

### (b) Inductor

n Symbol and model



$$v = L \frac{di}{dt}$$

$$i = \frac{1}{L} \int_{-\infty}^t v dt$$

$$= \frac{1}{L} \int_{-\infty}^0 v dt + \frac{1}{L} \int_0^t v dt$$

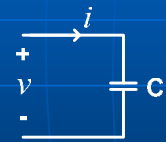
$$= i_0 + \frac{1}{L} \int_0^t v dt$$

L: inductance , in henry (H)

## 2.2 Passive Components

### (c) Capacitor

n Symbol and model



$$i = C \frac{dv}{dt}$$

$$v = \frac{1}{C} \int_{-\infty}^t i dt$$

$$= \frac{1}{C} \int_{-\infty}^0 i dt + \frac{1}{C} \int_0^t i dt$$

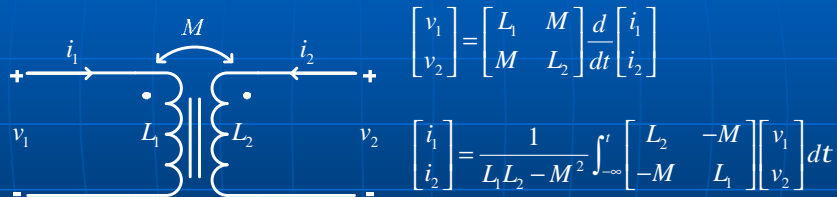
$$= v_0 + \frac{1}{C} \int_0^t i dt$$

C: capacitance , in Farad (F)

## 2.2 Passive Components

### (d) Coupling Inductor

n Symbol and model



$M$ : mutual inductance , in henry (H)

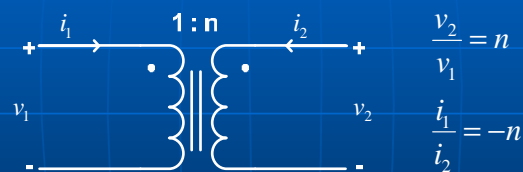
$L_1$ : primary self inductance , in henry (H)

$L_2$ : secondary self inductance , in henry (H)

## 2.2 Passive Components

### (e) Ideal Transformer

n Symbol and model



$n$ : turn ratio , positive real number

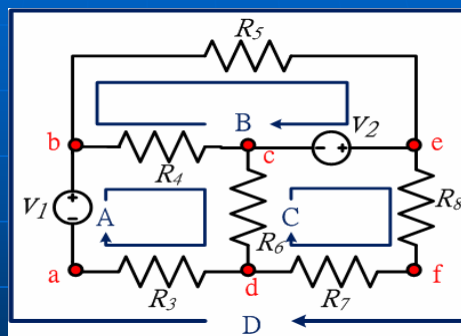
n It is convenient for modeling practical transformers or coupling inductors.

## 2.3 Node , Branch and Loop

- n A branch represents a single two-terminal element.
- n A node is a point where two or more elements join.
- n A loop is a closed path in a circuit without passing through any intermediate node more than once.

## 2.3 Node , Branch and Loop

n Example



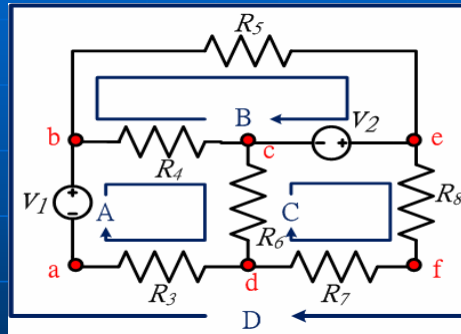
node : a , b , c , d , e , f      6 nodes

branch :  $V_1$  ,  $V_2$  ,  $R_3$  ,  $R_4$  ,  $R_5$  ,  $R_6$  ,  $R_7$  ,  $R_8$       8 branches



## 2.3 Node , Branch and Loop

n Example



loop A : a => b => c => d => a

loop B : b => e => c => b

loop C : c => e => f => d => c

loop D : a => b => e => f => d => a

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17

## 2.4 Kirchhoff's Laws

n Kirchhoff's current law (KCL)

The algebraic sum of all the current leaving any node in a circuit equals zero

$$\sum_{n=1}^N i_n = 0, \text{ for any node}$$

N : number of elements connected to this node

$i_n$  : the  $n$ th element current leaving this node

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18

## 2.4 Kirchhoff's Laws

- n The algebraic sum of all the current entering any node in a circuit equals zero

$$\sum_{n=1}^N (-i_n) = 0, \text{ for any node}$$

- n If  $N = L \cup M$ , then

$$\sum_L i_n = \sum_M i_m, \text{ for any node}$$

$L$ : set of  $n$  with  $i_n$  leaving this node

$M$ : set of  $m$  with  $i_m$  entering this node

## 2.4 Kirchhoff's Laws

- n KCL is based on the law of conservation of charge.
- n KCL can be applied to a super node (closed boundary)

## 2.4 Kirchhoff's Laws

n Kirchhoff's voltage law (KVL)

The algebraic sum of all the element voltage drops around any loop in a circuit equals zero.

$$\sum_{m=1}^M V_m = 0, \text{ for any loop}$$

## 2.4 Kirchhoff's Laws

n The algebraic sum of all the element voltage rises around any loop in a circuit equals zero.

$$\sum_{m=1}^M (-V_m) = 0, \text{ for any loop}$$

## 2.4 Kirchhoff's Laws

n If  $M = K + J$ , then

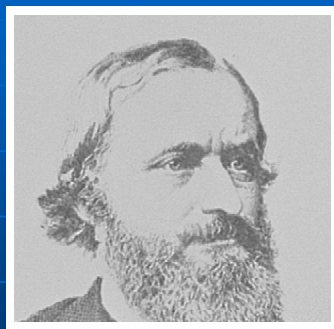
$$\sum_K i_k = \sum_J i_j$$

$M$  = set of  $k$  with  $V_k$  being a voltage drop in this loop

$J$  = set of  $j$  with  $V_j$  being a voltage rise in this loop

n KVL is based on the law of conservation of energy.

## 2.4 Kirchhoff's Laws



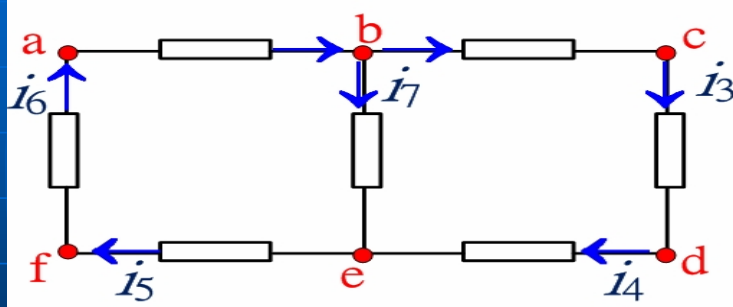
Gustav Robert Kirchhoff  
1824-1887  
Germany



???  
19XX-20XX  
R.O.C. Taiwan

## 2.4 Kirchhoff's Laws

n Example 1



$$\sum_k i_k (\text{leaving}) = 0, \text{ KCL in time domain.}$$

$$\text{Node b : } (-i_1) + i_7 + i_2 = 0$$

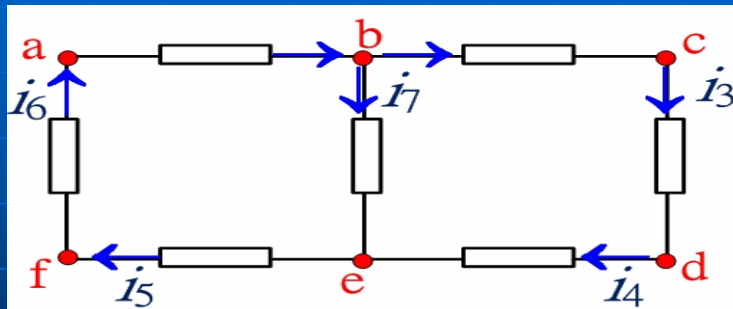
$$\text{Node e : } (-i_7) + (-i_4) + i_5 = 0$$

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25

## 2.4 Kirchhoff's Laws

n Example 1 (cont.)



$$\sum_j i_j (\text{entering}) = 0$$

$$\text{Node b : } i_1 + (-i_7) + (-i_2) = 0$$

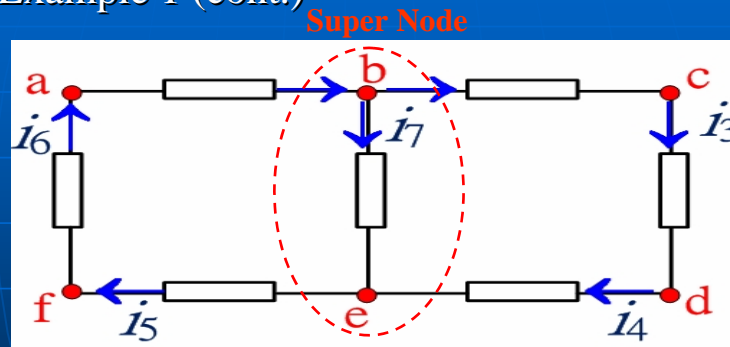
$$\text{Node e : } i_7 + i_4 + (-i_5) = 0$$

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26

## 2.4 Kirchhoff's Laws

n Example 1 (cont.)



$$\sum_m i_m(\text{leaving}) = \sum_n i_n(\text{entering})$$

$$\text{Node b : } i_7 + i_2 = i_1$$

$$\text{Node e : } i_5 = i_7 + i_4$$

$$\text{Super Node : } i_5 + i_2 = i_1 + i_4$$

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27

## 2.4 Kirchhoff's Laws

n Example 2

$$\sum_{n=1}^N i_n(t) = 0, \text{ for any node}$$

Assume

$$i_n(t) = I_n \cos(\omega t + f_n)$$

$$= \text{Re}[I_n e^{j(\omega t + f_n)}]$$

$$= \text{Re}[I_n e^{jf_n} e^{j\omega t}] \quad , \quad n=1,2,3,\dots,N$$

Re: real part

Define

$$\underline{I}_n = I_n e^{jf_n} = I_n \cos f_n + j I_n \sin f_n$$

Then

$$i_n(t) = \text{Re}[\underline{I}_n e^{j\omega t}]$$

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28

## 2.4 Kirchhoff's Laws

n Example 2 (cont.)

$$\sum_{n=1}^N \operatorname{Re}[I_n e^{j\omega t}] = 0$$

Q Re operator &  $\sum$  operator  
are commutative.

$$\operatorname{Re}\left[\sum_{n=1}^N I_n e^{j\omega t}\right] = 0$$

$$\therefore \sum_{n=1}^N I_n = 0, \text{ complex equation}$$

*KCL in phasor domain*

## 2.4 Kirchhoff's Laws

n Example 3

$$\sum_{n=1}^N i_n(t) = 0, \text{ for any node}$$

Let  $I_n(s) @ \int_0^{\infty} i_n(t) e^{-st} dt$

Then  $\int_0^{\infty} \sum_{n=1}^N i_n(t) dt = 0$

*$\int$  operator and  $\sum$  operator  
are commutative*

## 2.4 Kirchhoff's Laws

n Example 3 (cont.)

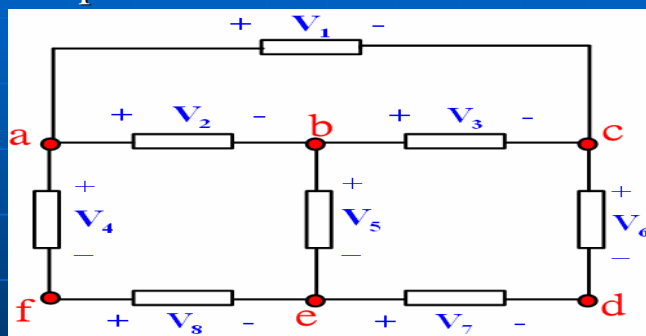
$$\sum_{n=1}^{\infty} \left( \int_0^{\infty} i_n(t) e^{-st} dt \right) = 0$$

$$\sum_{n=1}^{\infty} I_n(s) = 0, \text{ scalar equation}$$

*KCL in s domain*

## 2.4 Kirchhoff's Laws

n Example 4



$$\sum_K V_k(\text{drop}) = 0, \text{ KVL in time domain}$$

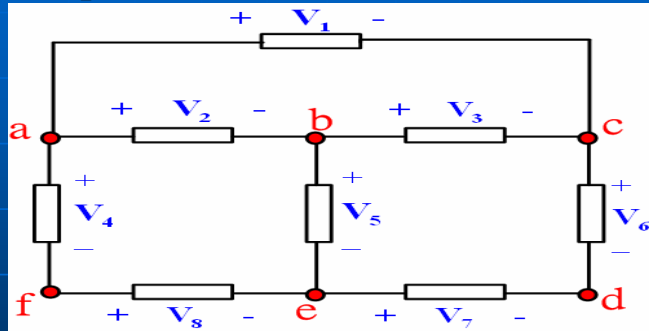
loop a, b, c, d, e, f, a

$$V_2 + V_3 + V_6 + (-V_7) + (-V_8) + (-V_4) = 0$$



## 2.4 Kirchhoff's Laws

n Example 4 (cont.)



$$\sum_K V_k (\text{rise}) = 0$$

loop a, b, c, d, e, f, a

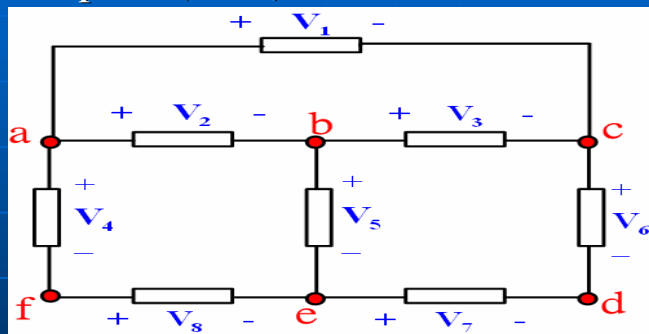
$$(-V_2) + (-V_3) + (-V_6) + V_7 + V_8 + V_4 = 0$$

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33

## 2.4 Kirchhoff's Laws

n Example 4 (cont.)



$$\sum V_n (\text{drop}) = \sum V_m (\text{rise})$$

loop a, b, c, d, e, f, a

$$V_2 + V_3 + V_6 = V_7 + V_8 + V_4$$

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34

## 2.4 Kirchhoff's Laws

n Example 5

$$\sum_{k=1}^M v_k(t) \text{ (drop)} = 0, \text{ for any node}$$

Assume  $v_k(t) = V_k \cos(\omega t + f_k)$   
 $\therefore v_k(t) = \text{Re}[V_k e^{jf_k} e^{j\omega t}]$

Let  $\underline{V}_k = V_k e^{jf_k} = V_k \cos f_k + jV_k \sin f_k$

Then  $v_k(t) = \text{Re}[\underline{V}_k e^{j\omega t}]$

## 2.4 Kirchhoff's Laws

n Example 5 (cont.)

Then 
$$\sum_{k=1}^M \text{Re}[\underline{V}_k e^{j\omega t}] = 0$$

$$\Rightarrow \text{Re}\left[\left(\sum_{k=1}^M \underline{V}_k\right) e^{j\omega t}\right] = 0$$

$$\Rightarrow \sum_{k=1}^M \underline{V}_k = 0, \text{ complex equation}$$

*KVL in phasor domain*

## 2.4 Kirchhoff's Laws

Example 6

$$\sum_{n=1}^N v_k(t) = 0$$

Define  $V_K(s) = \int_0^{\infty} v_k(t) e^{-st} dt$

$$\int_0^{\infty} \sum_{k=1}^M v_k(t) e^{-st} dt = 0$$

$$\sum_{k=1}^M \int_0^{\infty} v_k(t) e^{-st} dt = 0$$

$$\sum_{k=1}^M V_K(s) = 0, \text{ scalar equation}$$

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KVL in  $s$  domain

37

## 2.4 Kirchhoff's Laws

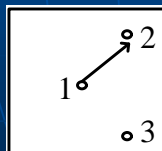
Example 7 (Chapter Problem 2.33)

Design an electric wiring system to control a single appliance from two or more locations.

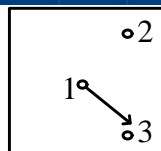
A three-way switch

: single pole double throw switch

(單極雙投開關)



Position 1



Position 2

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38

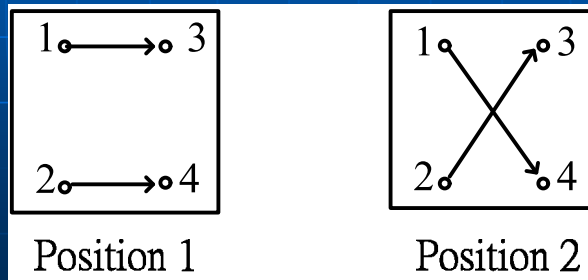
## 2.4 Kirchhoff's Laws

n Example 7 (cont.)

A four-way switch

: double pole double throw switch

(双極双投開關)



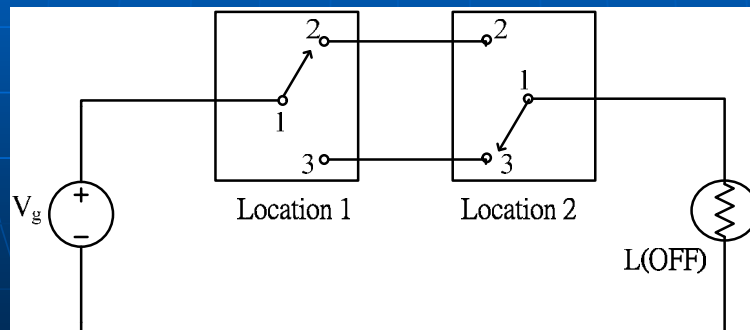
C.T. Pan

39

## 2.4 Kirchhoff's Laws

n Example 7 (cont.)

(a) Two three-way switches can be connected between two locations.



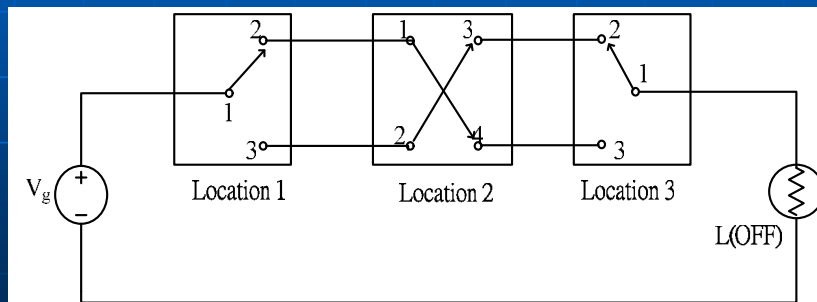
C.T. Pan

40

## 2.4 Kirchhoff's Laws

n Example 7 (cont.)

(b) One four-way switch is required for each location in excess of two.



## Summary

- n Objective 1 : Understand ideal circuit components.
- n Objective 2 : Be able to state and use component models.
- n Objective 3 : Be able to state and use KCL and KVL.

## Assignment : Chapter Problems

n Problem 2.3  
2.9  
2.26  
2.29

n Due within one week.

# THE END